1. (a) $\left[\begin{array}{ccc}0.1 & -0.3 & -0.4 \\ -0.5 & 0.1 & 0 \\ -0.2 & -0.3 & 0.4\end{array}\right]\left[\begin{array}{l}v_{1} \\ v_{2} \\ v_{3}\end{array}\right]=\left[\begin{array}{l}0 \\ 4 \\ 6\end{array}\right]$

Solving this matrix equation using a scientific calculator, $v_{2}=-8.387 \mathrm{~V}$
(b) Using a scientific calculator, the determinant is equal to 32.
2. (a) $\left[\begin{array}{rrr}1 & 1 & 1 \\ -1 & 2 & 3 \\ 2 & 0 & 4\end{array}\right]\left[\begin{array}{l}v_{\mathrm{A}} \\ v_{\mathrm{B}} \\ v_{\mathrm{C}}\end{array}\right]=\left[\begin{array}{c}27 \\ -16 \\ -6\end{array}\right]$

Solving this matrix equation using a scientific calculator,

$$
\begin{aligned}
& v_{\mathrm{A}}=19.57 \\
& v_{\mathrm{B}}=18.71 \\
& v_{\mathrm{C}}=-11.29
\end{aligned}
$$

(b) Using a scientific calculator,

$$
\left|\begin{array}{rrr}
1 & 1 & 1 \\
-1 & 2 & 3 \\
2 & 0 & 4
\end{array}\right|=16
$$

3. 

(a) We begin by simplifying the equations prior to solution:

$$
\begin{aligned}
4 & =0.08 v_{1}-0.05 v_{2}-0.02 v_{3} \\
8 & =-0.02 v_{1}-0.025 v_{2}+0.045 v_{3} \\
-2 & =-0.05 v_{1}+0.115 v_{2}-0.025 v_{3}
\end{aligned}
$$

Then, we can solve the matrix equation:

$$
\left[\begin{array}{ccc}
0.08 & -0.05 & -0.02 \\
-0.02 & -0.025 & 0.045 \\
-0.05 & 0.115 & -0.025
\end{array}\right]\left[\begin{array}{l}
v_{1} \\
v_{2} \\
v_{3}
\end{array}\right]=\left[\begin{array}{c}
4 \\
8 \\
-2
\end{array}\right]
$$

to obtain $v_{1}=264.3 \mathrm{~V}, v_{2}=183.9 \mathrm{~V}$ and $v_{3}=397.4 \mathrm{~V}$.
(b) We may solve the matrix equation directly using MATLAB, but a better check is to invoke the symbolic processor:

```
>> e1 = '4 = v1/100 + (v1 - v2)/20 + (v1 - vx)/50';
>> e2 = '10-4-(-2) = (vx - v1)/50 + (vx - v2)/40';
>> e3 = '-2 = v2/25 + (v2 - vx)/40 + (v2 - v1)/20';
>> a = solve(e1,e2,e3,'v1','v2','vx');
>> a.v1
ans =
82200/311
>> a.v2
ans =
57200/311
>> a.vx
ans =
```

4. We select the bottom node as our reference terminal and define two nodal voltages:


Ref.
Next, we write the two required nodal equations:
Node 1: $\quad 1=\frac{v_{1}}{2}+\frac{v_{1}-v_{2}}{3}$
Node 2: $\quad-3=\frac{v_{2}}{1}+\frac{v_{2}-v_{1}}{3}$
Which may be simplified to:

$$
\begin{aligned}
& 5 v_{1}-2 v_{2}=6 \\
& -v_{1}+4 v_{2}=-9
\end{aligned}
$$

and
Solving, we find that $v_{1}=333.3 \mathrm{mV}$.
5. We begin by selecting the bottom node as the reference terminal, and defining two nodal voltages $\mathrm{V}_{\mathrm{A}}$ and $\mathrm{V}_{\mathrm{B}}$, as shown. (Note if we choose the upper right node, $v_{1}$ becomes a nodal voltage and falls directly out of the solution.)


Ref.

We note that after completing nodal analysis, we can find $v_{1}$ as $v_{1}=V_{A}-V_{B}$.
At node $\mathrm{A}: ~ 4=\frac{\mathrm{V}_{\mathrm{A}}}{10}+\frac{\mathrm{V}_{\mathrm{A}}-\mathrm{V}_{\mathrm{B}}}{5}$
At node $\mathrm{B}:-(-6)=\frac{\mathrm{V}_{\mathrm{B}}}{8}+\frac{\mathrm{V}_{\mathrm{B}}-\mathrm{V}_{\mathrm{A}}}{5}$
Simplifying,

$$
\begin{aligned}
3 \mathrm{~V}_{\mathrm{A}}-2 \mathrm{~V}_{\mathrm{B}} & =40 \quad[1] \\
-8 \mathrm{~V}_{\mathrm{A}}+13 \mathrm{~V}_{\mathrm{B}} & =240 \quad[2]
\end{aligned}
$$

Solving, $\mathrm{V}_{\mathrm{A}}=43.48 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{B}}=45.22 \mathrm{~V}$, so $\mathrm{V}_{1}=-1.740 \mathrm{~V}$.
6. By inspection, no current flows through the $2 \Omega$ resistor, so $i_{1}=0$.

We next designate the bottom node as the reference terminal, and define $V_{A}$ and $V_{B}$ as shown:


At node $\mathrm{A}: 2=\frac{\mathrm{V}_{\mathrm{A}}}{3}+\frac{\mathrm{V}_{\mathrm{A}}-\mathrm{V}_{\mathrm{B}}}{1}$
At node B : $-2=\frac{\mathrm{V}_{\mathrm{B}}}{6}+\frac{\mathrm{V}_{\mathrm{B}}}{6}+\frac{\mathrm{V}_{\mathrm{B}}-\mathrm{V}_{\mathrm{A}}}{1}$
Note this yields $\mathrm{V}_{\mathrm{A}}$ and $\mathrm{V}_{\mathrm{B}}$, not $v_{1}$, due to our choice of reference node. So, we obtain $v_{1}$ by KVL: $v_{1}=\mathrm{V}_{\mathrm{A}}-\mathrm{V}_{\mathrm{B}}$.

Simplifying Eqs. [1] and [2],

$$
\begin{gather*}
4 \mathrm{~V}_{\mathrm{A}}-3 \mathrm{~V}_{\mathrm{B}}=6  \tag{1}\\
-3 \mathrm{~V}_{\mathrm{A}}+4 \mathrm{~V}_{\mathrm{B}}=-6 \tag{2}
\end{gather*}
$$

Solving, $\mathrm{V}_{\mathrm{A}}=0.8571 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{B}}=-0.8571 \mathrm{~V}$, so $v_{1}=1.714 \mathrm{~V}$.
7. The bottom node has the largest number of branch connections, so we choose that as our reference node. This also makes $v_{\mathrm{P}}$ easier to find, as it will be a nodal voltage. Working from left to right, we name our nodes $1, \mathrm{P}, 2$, and 3.

NODE 1: $\quad 10=v_{1} / 20+\left(v_{1}-v_{\mathrm{P}}\right) / 40$
NODE P: $\quad 0=\left(v_{\mathrm{P}}-v_{1}\right) / 40+v_{\mathrm{P}} / 100+\left(v_{\mathrm{P}}-v_{2}\right) / 50$
NODE 2: $\quad-2.5+2=\left(v_{2}-v_{\mathrm{P}}\right) / 50+\left(v_{2}-v_{3}\right) / 10$
NODE 3: $\quad 5-2=v_{3} / 200+\left(v_{3}-v_{2}\right) / 10$
Simplifying,

$$
\begin{array}{rlrl}
60 v_{1}-20 v_{\mathrm{P}} & =8000 & {[1]} \\
-50 v_{1}+110 v_{\mathrm{P}}-40 v_{2} & & =0 & {[2]} \\
-v_{\mathrm{P}}+6 v_{2}-5 v_{3} & =-25 & & {[3]} \\
-200 v_{2}+210 v_{3} & =6000 & {[4]}
\end{array}
$$

Solving,

$$
v_{\mathrm{P}}=171.6 \mathrm{~V}
$$

8. The logical choice for a reference node is the bottom node, as then $v_{\mathrm{x}}$ will automatically become a nodal voltage.

NODE 1: $\quad 4=v_{1} / 100+\left(v_{1}-v_{2}\right) / 20+\left(v_{1}-v_{\mathrm{x}}\right) / 50$
NODE $x: \quad 10-4-(-2)=\left(v_{x}-v_{1}\right) / 50+\left(v_{x}-v_{2}\right) / 40$
NODE 2: $-2=v_{2} / 25+\left(v_{2}-v_{\mathrm{x}}\right) / 40+\left(v_{2}-v_{1}\right) / 20$
Simplifying,

$$
\begin{array}{rll}
4 & =0.0800 v_{1}-0.0500 v_{2}-0.0200 v_{\mathrm{x}} & {[1]} \\
8 & =-0.0200 v_{1}-0.02500 v_{2}+0.04500 v_{\mathrm{x}} & {[2]} \\
-2 & =-0.0500 v_{1}+0.1150 v_{2}-0.02500 v_{\mathrm{x}} & {[3]}
\end{array}
$$

Solving,

$$
v_{x}=397.4 \mathrm{~V}
$$

9. Designate the node between the $3-\Omega$ and $6-\Omega$ resistors as node $X$, and the right-hand node of the $6-\Omega$ resistor as node $Y$. The bottom node is chosen as the reference node.
(a) Writing the two nodal equations, then

NODE X: $\quad-10=\left(v_{\mathrm{X}}-240\right) / 3+\left(v_{\mathrm{X}}-v_{\mathrm{Y}}\right) / 6$ [1]
NODE Y: $\quad 0=\left(v_{\mathrm{Y}}-v_{\mathrm{X}}\right) / 6+v_{\mathrm{Y}} / 30+\left(v_{\mathrm{Y}}-60\right) / 12$ [2]
Simplifying, $\quad-180+1440=9 v_{\mathrm{X}}-3 v_{\mathrm{Y}} \quad$ [1]

$$
10800=-360 v_{X}+612 v_{Y}
$$

Solving, $\quad v_{\mathrm{X}}=181.5 \mathrm{~V}$ and $\quad v_{\mathrm{Y}}=124.4 \mathrm{~V}$
Thus, $v_{1}=240-v_{\mathrm{X}}=58.50 \mathrm{~V}$ and $v_{2}=v_{\mathrm{Y}}-60=64.40 \mathrm{~V}$
(b) The power absorbed by the $6-\Omega$ resistor is

$$
\left(v_{\mathrm{X}}-v_{\mathrm{Y}}\right)^{2} / 6=543.4 \mathrm{~W}
$$

10. Only one nodal equation is required: At the node where three resistors join,

$$
\begin{equation*}
0.02 v_{1}=\left(v_{x}-5 i_{2}\right) / 45+\left(v_{x}-100\right) / 30+\left(v_{x}-0.2 v_{3}\right) / 50 \tag{1}
\end{equation*}
$$

This, however, is one equation in four unknowns, the other three resulting from the presence of the dependent sources. Thus, we require three additional equations:

$$
\begin{align*}
& i_{2}=\left(0.2 v_{3}-v_{\mathrm{x}}\right) / 50  \tag{2}\\
& v_{1}=0.2 v_{3}-100  \tag{3}\\
& v_{3}=50 i_{2} \tag{4}
\end{align*}
$$

Simplifying,

$$
\begin{array}{rlll}
v_{1}-0.2 v_{3} & =-100 & {[3]} \\
-v_{3}+50 i_{2} & =0 & {[4]} \\
+0.2 v_{3}-50 i_{2} & =0 & {[2]} \\
-v_{\mathrm{x}} & =0 & & \\
0.07556 v_{\mathrm{x}}-0.02 v_{1}-0.004 v_{3}-0.111 i_{2} & =33.33 & {[1]}
\end{array}
$$

Solving, we find that $v_{1}=-103 . .8 \mathrm{~V}$ and $i_{2}=-377.4 \mathrm{~mA}$.
11. If $v_{1}=0$, the dependent source is a short circuit and we may redraw the circuit as:


At NODE 1:

$$
4-6=v_{1} / 40+\left(v_{1}-96\right) / 20+\left(v_{1}-V_{2}\right) / 10
$$

Since $v_{1}=0$, this simplifies to

$$
-2=-96 / 20-V_{2} / 10
$$

so that $\mathrm{V}_{2}=-28 \mathrm{~V}$.
12. We choose the bottom node as ground to make calculation of $i_{5}$ easier. The left-most node is named " 1 ", the top node is named " 2 ", the central node is named " 3 " and the node between the $4-\Omega$ and $6-\Omega$ resistors is named "4."

NODE 1: $\quad-3=v_{1} / 2+\left(v_{1}-v_{2}\right) / 1$
NODE 2: $2=\left(v_{2}-v_{1}\right) / 1+\left(v_{2}-v_{3}\right) / 3+\left(v_{2}-v_{4}\right) / 4 \quad$ [2]
NODE 3: $3=v_{3} / 5+\left(v_{3}-v_{4}\right) / 7+\left(v_{3}-v_{2}\right) / 3$ [3]
NODE 4: $\quad 0=v_{4} / 6+\left(v_{4}-v_{3}\right) / 7+\left(v_{4}-v_{2}\right) / 4$ [4]
Rearranging and grouping terms,

$$
\begin{array}{rlll}
3 v_{1}-2 v_{2} & =-6 & {[1]} \\
-12 v_{1}+19 v_{2}-4 v_{3}-3 v_{4} & =24 & {[2]} \\
-35 v_{2}+71 v_{3}-15 v_{4} & =315 \\
-42 v_{2}-24 v_{3}+94 v_{4} & =0 & {[4]}
\end{array}
$$

Solving, we find that $v_{3}=6.760 \mathrm{~V}$ and so $i_{5}=v_{3} / 5=1.352 \mathrm{~A}$.
13. We can redraw this circuit and eliminate the $2.2-\mathrm{k} \Omega$ resistor as no current flows through it:


At NODE 2: $7 \times 10^{-3}-5 \times 10^{-3}=\left(v_{2}+9\right) / 470+\left(v_{2}-v_{\mathrm{x}}\right) / 10 \times 10^{-3}$
At NODE $x: \quad 5 \times 10^{-3}-0.2 v_{1}=\left(v_{x}-v_{2}\right) / 10 \times 10^{3}$
The additional equation required by the presence of the dependent source and the fact that its controlling variable is not one of the nodal voltages:

$$
\begin{equation*}
v_{1}=v_{2}-v_{\mathrm{x}} \tag{3}
\end{equation*}
$$

Eliminating the variable v1 and grouping terms, we obtain:

$$
10,470 v_{2}-470 v_{x}=-89,518
$$

and

$$
1999 v_{2}-1999 v_{\mathrm{x}}=50
$$

Solving, we find $\quad v_{x}=-8.086 \mathrm{~V}$.
14. We need concern ourselves with the bottom part of this circuit only. Writing a single nodal equation,

$$
-4+2=v / 50
$$

We find that $v=-100 \mathrm{~V}$.
15. We choose the bottom node as the reference terminal. Then:

Node 1: $\quad-2=\frac{v_{1}}{2}+\frac{v_{1}-v_{2}}{1}$
Node 2: $\quad 4=\frac{v_{2}-v_{1}}{1}+\frac{v_{2}-v_{3}}{2}+\frac{v_{2}-v_{4}}{4}$
Node 3: $\quad 2=\frac{v_{3}-v_{2}}{2}+\frac{v_{3}}{5}+\frac{v_{3}-v_{4}}{10}$
Node 4: $\quad 0=\frac{v_{4}}{6}+\frac{v_{4}-v_{3}}{10}+\frac{v_{4}-v_{2}}{4}$

Node 5: $\quad-1=\frac{v_{5}}{2}+\frac{v_{5}-v_{7}}{1}$
Node 6: $\quad 1=\frac{v_{6}}{5}+\frac{v_{6}-v_{7}}{2}+\frac{v_{6}-v_{8}}{10}$
Node 7: $\quad 2=\frac{v_{7}-v_{5}}{1}+\frac{v_{7}-v_{6}}{2}+\frac{v_{7}-v_{8}}{4}$
Node 8: $\quad 0=\frac{v_{8}}{6}+\frac{v_{8}-v_{6}}{10}+\frac{v_{8}-v_{7}}{4}$
Note that Eqs. [1-4] may be solved independently of Eqs. [5-8].
Simplifying,

| $3 v_{1}$ | $-2 v_{2}$ |  | $=-4$ | $[1]$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $-4 v_{1}$ | $+7 v_{2}$ | $-2 v_{3}$ | $-v_{4}$ | $=16$ | $[2]$ |
| $-5 v_{2}$ | $+8 v_{3}$ | $-v_{4}$ | $=20$ | $[3]$ |  |
| $-15 v_{2}$ | $-6 v_{3}$ | $+31 v_{4}$ | $=0$ | $[4]$ |  | to yield $\quad$| $v_{1}=3.370 \mathrm{~V}$ |
| :--- |
| $v_{2}=7.055 \mathrm{~V}$ |
| $v_{3}=7.518 \mathrm{~V}$ |
| $v_{4}=4.869 \mathrm{~V}$ |

$$
\begin{array}{cccccc}
\begin{array}{ccc}
\text { and } \\
3 v_{5}
\end{array} & & -2 v_{7} & & =-2 & {[5]} \\
& 8 v_{6} & -5 v_{7} & -v_{8} & =10 & {[6]} \\
-4 v_{5} & -2 v_{6} & +7 v_{7} & -v_{8} & =8 & {[7]} \\
& -6 v_{6} & -15 v_{7} & +31 v_{8} & =0 & {[8]}
\end{array} \text { to yield } \quad \begin{aligned}
& v_{5}=1.685 \mathrm{~V} \\
& v_{6}=3.759 \mathrm{~V} \\
& v_{7}=3.527 \mathrm{~V} \\
& v_{8}=2.434 \mathrm{~V} \\
& \hline
\end{aligned}
$$

16. We choose the center node for our common terminal, since it connects to the largest number of branches. We name the left node "A", the top node "B", the right node "C", and the bottom node "D". We next form a supernode between nodes A and B.

At the supernode: $\quad 5=\left(\mathrm{V}_{\mathrm{A}}-\mathrm{V}_{\mathrm{D}}\right) / 10+\mathrm{V}_{\mathrm{A}} / 20+\left(\mathrm{V}_{\mathrm{B}}-\mathrm{V}_{\mathrm{C}}\right) / 12.5$
At node $\mathrm{C}: \quad \mathrm{V}_{\mathrm{C}}=150$
At node D: $\quad-10=\mathrm{V}_{\mathrm{D}} / 25+\left(\mathrm{V}_{\mathrm{D}}-\mathrm{V}_{\mathrm{A}}\right) / 10$
Our supernode-related equation is $\mathrm{V}_{\mathrm{B}}-\mathrm{V}_{\mathrm{A}}=100$
Simplifiying and grouping terms,

$$
\begin{array}{rlll}
0.15 \mathrm{~V}_{\mathrm{A}}+0.08 \mathrm{~V}_{\mathrm{B}}-0.08 \mathrm{~V}_{\mathrm{C}}-0.1 \mathrm{~V}_{\mathrm{D}} & =5 & {[1]} \\
\mathrm{V}_{\mathrm{C}} & & =150 & {[2]} \\
-25 \mathrm{~V}_{\mathrm{A}} & +35 \mathrm{~V}_{\mathrm{D}} & =-2500 \\
-\mathrm{V}_{\mathrm{A}} & =\mathrm{V}_{\mathrm{B}} & & =100
\end{array}
$$

Solving, we find that $V_{D}=-63.06 \mathrm{~V}$. Since $v_{4}=-V_{D}$,

$$
v_{4}=63.06 \mathrm{~V}
$$

17. Choosing the bottom node as the reference terminal and naming the left node " 1 ", the center node " 2 " and the right node " 3 ", we next form a supernode about nodes 1 and 2 , encompassing the dependent voltage source.

At the supernode, At node 2,

$$
\begin{equation*}
5-8=\left(v_{1}-v_{2}\right) / 2+v_{3} / 2.5 \tag{1}
\end{equation*}
$$

Our
Our supernode equation is $v_{1}-v_{3}=0.8 v_{\mathrm{A}}$
Since $v_{\mathrm{A}}=v_{2}$, we can rewrite [3] as $v_{1}-v_{3}=0.8 v_{2}$
Simplifying and collecting terms,

$$
\begin{aligned}
0.5 v_{1}-0.5 v_{2}+0.4 v_{3} & =-3 \\
-0.5 v_{1}+0.7 v_{2} & =8 \\
v_{1}-0.8 v_{2}-v_{3} & =0
\end{aligned}
$$

(a) Solving for $v_{2}=v_{\mathrm{A}}$, we find that
$v_{\mathrm{A}}=25.91 \mathrm{~V}$
(b) The power absorbed by the $2.5-\Omega$ resistor is

$$
\left(v_{3}\right)^{2} / 2.5=(-0.4546)^{2} / 2.5 \quad=82.66 \mathrm{~mW}
$$

18. Selecting the bottom node as the reference terminal, we name the left node " 1 ", the middle node " 2 " and the right node " 3 ."

NODE 1: $\quad 5=\left(v_{1}-v_{2}\right) / 20+\left(v_{1}-v_{3}\right) / 50$
NODE 2: $\quad v_{2}=0.4 v_{1}$
NODE 3: $\quad 0.01 v_{1}=\left(v_{3}-v_{2}\right) / 30+\left(v_{3}-v_{1}\right) / 50$
Simplifying and collecting terms, we obtain

$$
\begin{array}{rrrr}
0.07 v_{1}-0.05 v_{2}-0.02 v_{3} & =5 & {[1]} \\
0.4 v_{1}-1-v_{2} & =0 & {[2]} \\
-0.03 v_{1}-0.03333 v_{2}+0.05333 v_{3} & =0 & {[3]}
\end{array}
$$

Since our choice of reference terminal makes the controlling variable of both dependent sources a nodal voltage, we have no need for an additional equation as we might have expected.
Solving, we find that $v_{1}=148.2 \mathrm{~V}, v_{2}=59.26 \mathrm{~V}$, and $v_{3}=120.4 \mathrm{~V}$.
The power supplied by the dependent current source is therefore

$$
\left(0.01 v_{1}\right) \cdot v_{3}=177.4 \mathrm{~W}
$$

19. At node x: $v_{\mathrm{x}} / 4+\left(v_{\mathrm{x}}-v_{\mathrm{y}}\right) / 2+\left(v_{\mathrm{x}}-6\right) / 1 \quad=0$ [1]

At node $\mathrm{y}: \quad\left(v_{\mathrm{y}}-k v_{\mathrm{x}}\right) / 3+\left(v_{\mathrm{y}}-v_{\mathrm{x}}\right) / 2 \quad=2 \quad$ [2]
Our additional constraint is that $v_{\mathrm{y}}=0$, so we may simplify Eqs. [1] and [2]:

$$
\begin{align*}
& 14 v_{\mathrm{x}}=48  \tag{1}\\
& -2 \mathrm{k} v_{\mathrm{x}}-3 v_{\mathrm{x}}=12
\end{align*}
$$

Since Eq. [1] yields $v_{\mathrm{x}}=48 / 14=3.429 \mathrm{~V}$, we find that

$$
k=\left(12+3 v_{\mathrm{x}}\right) /\left(-2 v_{\mathrm{x}}\right)=-3.250
$$

20. Choosing the bottom node joining the $4-\Omega$ resistor, the $2-\mathrm{A}$ current sourcee and the $4-\mathrm{V}$ voltage source as our reference node, we next name the other node of the $4-\Omega$ resistor node " 1 ", and the node joining the $2-\Omega$ resistor and the 2 -A current source node "2." Finally, we create a supernode with nodes " 1 " and " 2. ."

At the supernode:

$$
\begin{align*}
& -2=v_{1} / 4+\left(v_{2}-4\right) / 2  \tag{1}\\
& v_{1}-v_{2}=-3-0.5 i_{1}  \tag{2}\\
& i_{1}=\left(v_{2}-4\right) / 2 \tag{3}
\end{align*}
$$

Our remaining equations:
and
Equation [1] simplifies to $\quad v_{1}+2 v_{2}=0 \quad$ [1]
Combining Eqs. [2] and [3, $4 v_{1}-3 v_{2}=-8 \quad$ [4]
Solving these last two equations, we find that $v_{2}=727.3 \mathrm{mV}$. Making use of Eq. [3], we therefore find that

$$
i_{1}=-1.636 \mathrm{~A} .
$$

21. We first number the nodes as $1,2,3,4$, and 5 moving left to right. We next select node 5 as the reference terminal. To simplify the analysis, we form a supernode from nodes 1,2 , and 3 .

At the supernode,

$$
\begin{equation*}
-4-8+6=v_{1} / 40+\left(v_{1}-v_{3}\right) / 10+\left(v_{3}-v_{1}\right) / 10+v_{2} / 50+\left(v_{3}-v_{4}\right) / 20 \tag{1}
\end{equation*}
$$

Note that since both ends of the $10-\Omega$ resistor are connected to the supernode, the related terms cancel each other out, and so could have been ignored.

At node 4:

$$
\begin{equation*}
v_{4}=200 \tag{2}
\end{equation*}
$$

Supernode KVL equation:

$$
\begin{equation*}
v_{1}-v_{3}=400+4 v_{20} \tag{3}
\end{equation*}
$$

Where the controlling voltage

$$
\begin{equation*}
v_{20}=v_{3}-v_{4}=v_{3}-200 \tag{4}
\end{equation*}
$$

Thus, Eq. [1] becomes $-6=v_{1} / 40+v_{2} / 50+\left(v_{3}-200\right) / 20$ or, more simply,

$$
4=v_{1} / 40+v_{2} / 50+v_{3} / 20 \quad\left[1^{\prime}\right]
$$

and Eq. [3] becomes

$$
\begin{equation*}
v_{1}-5 v_{3}=-400 \tag{3’}
\end{equation*}
$$

Eqs. [1'], [3'], and [5] are not sufficient, however, as we have four unknowns. At this point we need to seek an additional equation, possibly in terms of $v_{2}$. Referring to the circuit,

$$
\begin{equation*}
v_{1}-v_{2}=400 \tag{5}
\end{equation*}
$$

Rewriting as a matrix equation,

$$
\left[\begin{array}{ccc}
1 / 40 & 1 / 50 & 1 / 20 \\
1 & 0 & -5 \\
1 & -1 & 0
\end{array}\right]\left[\begin{array}{l}
v_{1} \\
v_{2} \\
v_{3}
\end{array}\right]=\left[\begin{array}{c}
4 \\
-400 \\
400
\end{array}\right]
$$

Solving, we find that
$v_{1}=145.5 \mathrm{~V}, v_{2}=-254.5 \mathrm{~V}$, and $v_{3}=109.1 \mathrm{~V}$. Since $v_{20}=v_{3}-200$, we find that

$$
V_{20}=-90.9 \mathrm{~V} .
$$

22. We begin by naming the top left node " 1 ", the top right node " 2 ", the bottom node of the $6-\mathrm{V}$ source " 3 " and the top node of the $2-\Omega$ resistor " 4 ." The reference node has already been selected, and designated using a ground symbol.

By inspection, $v_{2}=5 \mathrm{~V}$.
Forming a supernode with nodes $1 \& 3$, we find
At the supernode:

$$
\begin{equation*}
-2=v_{3} / 1+\left(v_{1}-5\right) / 10 \tag{1}
\end{equation*}
$$

At node 4:

$$
\begin{equation*}
2=v_{4} / 2+\left(v_{4}-5\right) / 4 \tag{2}
\end{equation*}
$$

Our supernode KVL equation: $\quad v_{1}-v_{3}=6$
Rearranging, simplifying and collecting terms,

$$
\begin{equation*}
v_{1}+10 v_{3}=-20+5=-15 \tag{1}
\end{equation*}
$$

and

$$
\begin{equation*}
v_{1}-v_{3}=6 \tag{2}
\end{equation*}
$$

Eq. [3] may be directly solved to obtain $v_{4}=4.333 \mathrm{~V}$.
Solving Eqs. [1] and [2], we find that

$$
v_{1}=4.091 \mathrm{~V} \text { and } v_{3}=-1.909 \mathrm{~V} .
$$

23. We begin by selecting the bottom node as the reference, naming the nodes as shown below, and forming a supernode with nodes 5 \& 6 .


By inspection, $\quad v_{4}=4 \mathrm{~V}$.
By KVL, $v_{3}-v_{4}=1$ so $v_{3}=-1+v_{4}=-1+4$ or $v_{3}=3 \mathrm{~V}$.
At the supernode, $2=v_{6} / 1+\left(v_{5}-4\right) / 2 \quad$ [1]
At node 1, $4=v_{1} / 3 \quad$ therefore, $\quad v_{1}=12 \mathrm{~V}$.
At node 2, $\quad-4-2=\left(v_{2}-3\right) / 4$
Solving, we find that $\quad v_{2}=-21 \mathrm{~V}$
Our supernode KVL equation is $\quad v_{5}-v_{6}=3$ [2]
Solving Eqs. [1] and [2], we find that

$$
v_{5}=4.667 \mathrm{~V} \text { and } v_{6}=1.667 \mathrm{~V}
$$

The power supplied by the 2-A source therefore is $\left(v_{6}-v_{2}\right)(2)=45.33 \mathrm{~W}$.
24. We begin by selecting the bottom node as the reference, naming each node as shown below, and forming two different supernodes as indicated.


Voltages in volts. Currents in amperes. Resistances in ohms.

By inspection, $\quad v_{7}=4 \mathrm{~V}$ and $v_{1}=(3)(4)=12 \mathrm{~V}$.
At node 2:

$$
\begin{equation*}
-4-2=\left(v_{2}-v_{3}\right) / 4 \quad \text { or } \quad v_{2}-v_{3}=-24 \tag{1}
\end{equation*}
$$

At the 3-4 supernode:

$$
\begin{equation*}
0=\left(v_{3}-v_{2}\right) / 4+\left(v_{4}-v_{5}\right) / 6 \quad \text { or } \quad-6 v_{2}+6 v_{3}+4 v_{4}-4 v_{5}=0 \tag{2}
\end{equation*}
$$

At node 5:
$0=\left(v_{5}-v_{4}\right) / 6+\left(v_{5}-4\right) / 7+\left(v_{5}-v_{6}\right) / 2$ or $-14 v_{4}+68 v_{5}-42 v_{6}=48$
At the 6-8 supernode: $2=\left(v_{6}-v_{5}\right) / 2+v_{8} / 1 \quad$ or $\quad-v_{5}+v_{6}+2 v_{8}=4$
3-4 supernode KVL equation: $\quad v_{3}-v_{4}=-1$
$6-8$ supernode KVL equation: $\quad v_{6}-v_{8}=3$
Rewriting Eqs. [1] to [6] in matrix form,

$$
\left[\begin{array}{lccccc}
1 & -1 & 0 & 0 & 0 & 0 \\
-6 & 6 & 4 & -4 & 0 & 0 \\
0 & 0 & -14 & 68 & -42 & 0 \\
0 & 0 & 0 & -1 & 1 & 2 \\
0 & 1 & -1 & 0 & 0 & 0 \\
0 & 0 & 0 & 0 & 1 & -1
\end{array}\right]\left[\begin{array}{c}
v_{2} \\
v_{3} \\
v_{4} \\
v_{5} \\
v_{6} \\
v_{8}
\end{array}\right]=\left[\begin{array}{r}
-24 \\
0 \\
48 \\
4 \\
-1 \\
3
\end{array}\right]
$$

Solving, we find that

$$
v_{2}=-68.9 \mathrm{~V}, v_{3}=-44.9 \mathrm{~V}, v_{4}=-43.9 \mathrm{~V}, v_{5}=-7.9 \mathrm{~V}, v_{6}=700 \mathrm{mV}, v_{8}=-2.3 \mathrm{~V} .
$$

The power generated by the 2-A source is therefore $\left(v_{8}-v_{6}\right)(2)=133.2 \mathrm{~W}$.
25. With the reference terminal already specified, we name the bottom terminal of the $3-\mathrm{mA}$ source node " 1 ," the left terminal of the bottom $2.2-\mathrm{k} \Omega$ resistor node " 2 ," the top terminal of the $3-\mathrm{mA}$ source node " 3 ," the " + " reference terminal of the $9-\mathrm{V}$ source node " 4 ," and the "-" terminal of the $9-\mathrm{V}$ source node " 5 ."

Since we know that 1 mA flows through the top 2.2-k resistor, $\quad v_{5}=-2.2 \mathrm{~V}$.
Also, we see that $v_{4}-v_{5}=9$, so that $v_{4}=9-2.2=6.8 \mathrm{~V}$.
Proceeding with nodal analysis,
At node 1: $\quad-3 \times 10^{-3}=v_{1} / 10 \times 10^{3}+\left(v_{1}-v_{2}\right) / 2.2 \times 10^{3}$
At node 2: $\quad 0=\left(v_{2}-v_{1}\right) / 2.2 \times 10^{3}+\left(v_{2}-v_{3}\right) / 4.7 \times 10^{3}$
At node 3: $\quad 1 \times 10^{3}+3 \times 10^{3}=\left(v_{3}-v_{2}\right) / 4.7 \times 10^{3}+v_{3} / 3.3 \times 10^{3}$
Solving,

$$
\begin{equation*}
v_{1}=-8.614 \mathrm{~V}, v_{2}=-3.909 \mathrm{~V} \text { and } v_{3}=6.143 \mathrm{~V} \tag{3}
\end{equation*}
$$

Note that we could also have made use of the supernode approach here.
26. Mesh 1: $-4+400 i_{1}+300 i_{1}-300 i_{2}-1=0$ or $700 i_{1}-300 i_{2}=5$

Mesh 2: $1+500 i_{2}-300 i_{1}+2-2=0 \quad$ or $-300 i_{1}+500 i_{2}=-3.2$
Solving, $i_{1}=5.923 \mathrm{~mA}$ and $i_{2}=-2.846 \mathrm{~mA}$.
27. (a) Define a clockwise mesh current $i_{1}$ in the left-most mesh; a clockwise mesh current $i_{2}$ in the central mesh, and note that $i_{y}$ can be used as a mesh current for the remaining mesh.

Mesh 1: $-10+7 i_{1}-2 i_{2}=0$
Mesh 2: $\quad-2 i_{1}+5 i_{2}=0$
Mesh $y: \quad-2 i_{2}+9 i_{y}=0$
Solve the resulting matrix equation:

$$
\left[\begin{array}{ccc}
7 & -2 & 0 \\
-2 & 5 & 0 \\
0 & -2 & 9
\end{array}\right]\left[\begin{array}{l}
i_{1} \\
i_{2} \\
i_{y}
\end{array}\right]=\left[\begin{array}{c}
10 \\
0 \\
0
\end{array}\right] \text { to find that } i_{1}=1.613 \mathrm{~A} \text {, and } i_{y}=143.4 \mathrm{~mA} .
$$

(b) The power supplied by the 10 V source is $(10)\left(i_{1}\right)=10(1.613)=16.13 \mathrm{~W}$.
28. Define three mesh currents as shown:

(a) The current through the $2 \Omega$ resistor is $i_{1}$.

Mesh 1: $5 i_{1}-3 i_{2}=0$
Mesh 2: $-212+8 i_{2}-3 i_{1}=0$
Mesh 3: $8 i_{3}-5 i_{2}+122=0$

$$
\begin{array}{ll}
\text { or } 5 i_{1}-3 i_{2} & = \\
\text { or }-3 i_{1}+8 i_{2} & = \\
\text { or }-5 i_{2}+8 i_{3} & =-122 \\
\text { or }
\end{array}
$$

Solving $i_{1}=20.52 \mathrm{~A}, i_{2}=34.19 \mathrm{~A}$ and $i_{3}=6.121 \mathrm{~A}$.
(b) The current through the $5 \Omega$ resistor is $i_{3}$, or 6.121 A.
*** Note: since the problem statement did not specify a direction, only the current magnitude is relevant, and its sign is arbitrary.
29. We begin by defining three clockwise mesh currents $i_{1}, i_{2}$ and $i_{3}$ in the left-most, central, and right-most meshes, respectively. Then,
(a) Note that $i_{x}=i_{2}-i_{3}$.

Mesh 1: $i_{1}=5 \mathrm{~A}$ (by inspection)
Mesh 3: $i_{3}=-2$ A (by inspection)
Mesh 2: $-25 i_{1}+75 i_{2}-20 i_{3}=0$, or, making use of the above,

$$
-125+75 i_{2}+40=0 \text { so that } i_{2}=1.133 \mathrm{~A} .
$$

Thus, $i_{x}=i_{2}-i_{3}=1.133-(-2)=3.133 \mathrm{~A}$.
(b) The power absorbed by the $25 \Omega$ resistor is

$$
\mathrm{P}_{25 \Omega}=25\left(i_{1}-i_{2}\right)^{2}=25(5-1.133)^{2}=373.8 \mathrm{~W} .
$$

30. Define three mesh currents as shown. Then,


Mesh 1

$$
-2+80 i_{1}-40 i_{2}-30 i_{3}=0
$$

Mesh 2: $\quad-40 i_{1}+70 i_{2} \quad=0$
Mesh 3: $\quad-30 i_{1} \quad+70 i_{3}=0$
Solving, $\left[\begin{array}{ccc}80 & -40 & -30 \\ -40 & 70 & 0 \\ -30 & 0 & 70\end{array}\right]\left[\begin{array}{l}i_{1} \\ i_{2} \\ i_{3}\end{array}\right]=\left[\begin{array}{l}2 \\ 0 \\ 0\end{array}\right]$
we find that $i_{2}=25.81 \mathrm{~mA}$ and $i_{3}=19.35 \mathrm{~mA}$. Thus, $i=i_{3}-i_{2}=-6.46 \mathrm{~mA}$.
31. Moving from left to right, we name the bottom three meshes, mesh " 1 ", mesh "2," and mesh " 3. ." In each of these three meshes we define a clockwise current. The remaining mesh current is clearly 8 A . We may then write:

MESH 1: $12 i_{1}-4 i_{2}=100$
MESH 2: $\quad-4 i_{1}+9 i_{2}-3 i_{3}=0$
MESH 3:
$-3 i_{2}+18 i_{3}=-80$
Solving this system of three (independent) equations in three unknowns, we find that

$$
i_{2}=i_{\mathrm{x}}=2.791 \mathrm{~A} .
$$

32. We define four clockwise mesh currents. The top mesh current is labeled $i_{4}$. The bottom left mesh current is labeled $i_{1}$, the bottom right mesh current is labeled $i_{3}$, and the remaining mesh current is labeled $i_{2}$. Define a voltage " $v_{4 \mathrm{~A}}$ " across the 4-A current source with the " + " reference terminal on the left.

By inspection, $\quad i_{3}=5 \mathrm{~A}$ and $i_{\mathrm{a}}=i_{4}$.
MESH 1: $-60+2 i_{1}-2 i_{4}+6 i_{4}=0 \quad$ or $2 i_{1}+4 i_{4}=60$
MESH 2: $-6 i_{4}+v_{4 \mathrm{~A}}+4 i_{2}-4(5)=0 \quad$ or $\quad 4 i_{2}-6 i_{4}+v_{4 \mathrm{~A}}=20$
MESH 4: $2 i_{4}-2 i_{1}+5 i_{4}+3 i_{4}-3(5)-v_{4 \mathrm{~A}}=0 \quad$ or $-2 i_{1}+10 i_{4}-v_{4 \mathrm{~A}}=15$
At this point, we are short an equation. Returning to the circuit diagram, we note that

$$
\begin{equation*}
i_{2}-i_{4}=4 \tag{4}
\end{equation*}
$$

Collecting these equations and writing in matrix form, we have

$$
\left[\begin{array}{rrrr}
2 & 0 & 4 & 0 \\
0 & 4 & -6 & 1 \\
-2 & 0 & 10 & -1 \\
0 & 1 & -1 & 0
\end{array}\right]\left[\begin{array}{l}
i_{1} \\
i_{2} \\
i_{4} \\
v_{4 \mathrm{~A}}
\end{array}\right]=\left[\begin{array}{c}
60 \\
20 \\
15 \\
4
\end{array}\right]
$$

Solving, $i_{1}=16.83 \mathrm{~A}, i_{2}=10.58 \mathrm{~A}, i_{4}=6.583 \mathrm{~A}$ and $v_{4 \mathrm{~A}}=17.17 \mathrm{~V}$.
Thus, the power dissipated by the $2-\Omega$ resistor is

$$
\left(i_{1}-i_{4}\right)^{2} \cdot(2)=210.0 \mathrm{~W}
$$

33. We begin our analysis by defining three clockwise mesh currents. We will call the top mesh current $i_{3}$, the bottom left mesh current $i_{1}$, and the bottom right mesh current $i_{2}$.

By inspection, $i_{1}=5 \mathrm{~A}[1] \quad$ and $i_{2}=-0.01 v_{1}$
MESH 3: $\quad 50 i_{3}+30 i_{3}-30 i_{2}+20 i_{3}-20 i_{1}=0$
or $\quad-20 i_{1}-30 i_{2}+100 i_{3}=0 \quad$ [3]
These three equations are insufficient, however, to solve for the unknowns. It would be nice to be able to express the dependent source controlling variable $v_{1}$ in terms of the mesh currents. Returning to the diagram, it can be seen that KVL around mesh 1 will yield

$$
-v_{1}+20 i_{1}-20 i_{3}+0.4 v_{1}=0
$$

or

$$
v_{1}=20 i_{1} / 0.6-20 i_{3} / 0.6 \quad \text { or } \quad v_{1}=\left(20(5) / 0.6-20 i_{3} / 0.6[4]\right.
$$

Substituting Eq. [4] into Eq. [2] and then the modified Eq. [2] into Eq. [3], we find

$$
-20(5)-30(-0.01)(20)(5) / 0.6+30(-0.01)(20) i_{3} / 0.6+100 i_{3}=0
$$

Solving, we find that $i_{3}=(100-50) / 90=555.6 \mathrm{~mA}$
Thus, $v_{1}=148.1 \mathrm{~V}, i_{2}=-1.481 \mathrm{~A}$, and the power generated by the dependent voltage source is

$$
0.4 v_{1}\left(i_{2}-i_{1}\right)=-383.9 \mathrm{~W} .
$$

34. We begin by defining four clockwise mesh currents $i_{1}, i_{2}, i_{3}$ and $i_{4}$, in the meshes of our circuit, starting at the left-most mesh. We also define a voltage $v_{\text {dep }}$ across the dependent current source, with the " + " on the top.
By inspection, $i_{1}=2 \mathrm{~A}$ and $i_{4}=-5 \mathrm{~A}$.
At Mesh 2: $\quad 10 i_{2}-10(2)+20 i_{2}+v_{\text {dep }}=0 \quad$ [1]
At Mesh 3: $\quad-v_{\text {dep }}+25 i_{3}+5 i_{3}-5(-5)=0 \quad$ [2]
Collecting terms, we rewrite Eqs. [1] and [2] as

$$
\begin{align*}
30 i_{2} & +v_{\text {dep }} \tag{1}
\end{align*}=20 .
$$

This is only two equations but three unknowns, however, so we require an additional equation. Returning to the circuit diagram, we note that it is possible to express the current of the dependent source in terms of mesh currents. (We might also choose to obtain an expression for $v_{\text {dep }}$ in terms of mesh currents using KVL around mesh 2 or 3.)

Thus, $1.5 i_{\mathrm{x}}=i_{3}-i_{2}$ where $i_{\mathrm{x}}=i_{1}-i_{2}$ so $-0.5 i_{2}-i_{3}=-3$
In matrix form,

$$
\left[\begin{array}{rrr}
30 & 0 & 1 \\
0 & 30 & -1 \\
-0.5 & -1 & 0
\end{array}\right]\left[\begin{array}{l}
i_{2} \\
i_{3} \\
v_{\text {dep }}
\end{array}\right]=\left[\begin{array}{r}
20 \\
-25 \\
-3
\end{array}\right]
$$

Solving, we find that $i_{2}=-6.333 \mathrm{~A}$ so that $i_{\mathrm{x}}=i_{1}-i_{2}=8.333 \mathrm{~A}$.
35. We define a clockwise mesh current $i_{1}$ in the bottom left mesh, a clockwise mesh current $i_{2}$ in the top left mesh, a clockwise mesh current $i_{3}$ in the top right mesh, and a clockwise mesh current $i_{4}$ in the bottom right mesh.

MESH 1: $\quad-0.1 v_{\mathrm{a}}+4700 i_{1}-4700 i_{2}+4700 i_{1}-4700 i_{4}=0$
MESH 2: $\quad 9400 i_{2}-4700 i_{1}-9=0$
MESH 3: $\quad 9+9400 i_{3}-4700 i_{4}=0$
MESH 4: $\quad 9400 i_{4}-4700 i_{1}-4700 i_{3}+0.1 i_{\mathrm{x}}=0$
The presence of the two dependent sources has led to the introduction of two additional unknowns ( $i_{\mathrm{x}}$ and $v_{\mathrm{a}}$ ) besides our four mesh currents. In a perfect world, it would simplify the solution if we could express these two quantities in terms of the mesh currents.

Referring to the circuit diagram, we see that $i_{\mathrm{x}}=i_{2}$ (easy enough) and that $v_{\mathrm{a}}=4700 i_{3} \quad$ (also straightforward). Thus, substituting these expressions into our four mesh equations and creating a matrix equation, we arrive at:

$$
\left[\begin{array}{rrrr}
9400 & -4700 & -470 & -4700 \\
-4700 & 9400 & 0 & 0 \\
0 & 0 & 9400 & -4700 \\
-4700 & 0.1 & -4700 & 9400
\end{array}\right]\left[\begin{array}{l}
i_{1} \\
i_{2} \\
i_{3} \\
i_{4}
\end{array}\right]=\left[\begin{array}{r}
0 \\
9 \\
-9 \\
0
\end{array}\right]
$$

Solving,

$$
i_{1}=239.3 \mu \mathrm{~A}, i_{2}=1.077 \mathrm{~mA}, i_{3}=-1.197 \mathrm{~mA} \text { and } i_{4}=-478.8 \mu \mathrm{~A} .
$$

36. We define a clockwise mesh current $i_{3}$ in the upper right mesh, a clockwise mesh current $i_{1}$ in the lower left mesh, and a clockwise mesh current $i_{2}$ in the lower right mesh.

MESH 1: $\quad-6+6 i_{1}-2=0$
MESH 2: $\quad 2+15 i_{2}-12 i_{3}-1.5=0$
MESH 3: $\quad i_{3}=0.1 v_{\mathrm{x}}$
Eq. [1] may be solved directly to obtain

$$
i_{1}=1.333 \mathrm{~A} .
$$

It would help in the solution of Eqs. [2] and [3] if we could express the dependent source controlling variable $v_{\mathrm{x}}$ in terms of mesh currents. Referring to the circuit diagram, we see that $v_{\mathrm{x}}=(1)\left(i_{1}\right)=i_{1}$, so Eq. [3] reduces to

$$
i_{3}=0.1 v_{\mathrm{x}}=0.1 i_{1}=133.3 \mathrm{~mA} .
$$

As a result, Eq. [1] reduces to

$$
i_{2}=[-0.5+12(0.1333)] / 15=73.31 \mathrm{~mA} .
$$

37. (a) Define a mesh current $i_{2}$ in the second mesh. Then KVL allows us to write:

MESH 1: $-9+R i_{1}+47000 i_{1}-47000 i_{2}=0 \quad[1]$
MESH 2: $67000 i_{2}-47000 i_{1}-5=0$
Given that $i_{1}=1.5 \mathrm{~mA}$, we may solve Eq. [2] to find that

$$
i_{2}=\frac{5+47(1.5)}{67} \mathrm{~mA}=1.127 \mathrm{~mA}
$$

and so

$$
R=\frac{9-47(1.5)+47(1.127)}{1.5 \times 10^{-3}}=-5687 \Omega
$$

(b) This value of $R$ is unique; no other value will satisfy both Eqs. [1] and [2].
38. Define three clockwise mesh currents $i_{1}, i_{2}$ and $i_{3}$. The bottom $1-\mathrm{k} \Omega$ resistor can be ignored, as no current flows through it.

MESH 1: $\quad-4+(2700+1000+5000) i_{1}-1000 i_{2}=0$
MESH 2: $\quad(1000+1000+4400+3000) i_{2}-1000 i_{1}-4400 i_{3}+2.2-3=0$
MESH 3: $\quad(4400+4000+3000) i_{3}-4400 i_{2}-1.5=0$
Combining terms,

$$
\begin{align*}
8700 i_{1}-1000 i_{2} & =4 \\
-1000 i_{1}+9400 i_{2}-4400 i_{3} & =0.8  \tag{2}\\
-4400 i_{2}+11400 i_{3} & =1.5 \tag{3}
\end{align*}
$$

Solving,

$$
i_{1}=487.6 \mu \mathrm{~A}, i_{2}=242.4 \mu \mathrm{~A} \text { and } i_{3}=225.1 \mu \mathrm{~A} .
$$

The power absorbed by each resistor may now be calculated:

| $\mathrm{P}_{5 \mathrm{k}}$ | $5000\left(i_{1}\right)^{2}$ | 1.189 mW |
| :---: | :---: | :---: |
| $\mathrm{P}_{2.7 \mathrm{k}}$ | $2700\left(i_{1}\right)^{2}$ | $641.9 \mu \mathrm{~W}$ |
| $\mathrm{P}_{1 \text { ktop }}$ | $1000\left(i_{1}-i_{2}\right)^{2}$ | $60.12 \mu \mathrm{~W}$ |
| $\mathrm{P}_{1 \mathrm{kmiddle}}=$ | $1000\left(i_{2}\right)^{2}$ | $58.76 \mu \mathrm{~W}$ |
| $\mathrm{P}_{1 \mathrm{kbottom}}=$ | 0 | 0 |
| $\mathrm{P}_{4.4 \mathrm{k}}$ | $4400\left(i_{2}-i_{3}\right)^{2}$ | $1.317 \mu \mathrm{~W}$ |
| $\mathrm{P}_{3 \text { ktop }}$ | $3000\left(i_{3}\right)^{2}$ | $152.0 \mu \mathrm{~W}$ |
| $\mathrm{P}_{4 \mathrm{k}}$ | $4000\left(i_{3}\right)^{2}$ | $202.7 \mu \mathrm{~W}$ |
| $\mathrm{P}_{3 \mathrm{kbottom}}=$ | $3000\left(i_{2}\right)^{2}$ | $176.3 \mu \mathrm{~W}$ |

Check: The sources supply a total of

$$
4(487.6)+(3-2.2)(242.4)+1.5(225.1)=2482 \mu \mathrm{~W} .
$$

The absorbed powers add to $2482 \mu \mathrm{~W}$.
39. (a) We begin by naming four mesh currents as depicted below:


Proceeding with mesh analysis, then, keeping in mind that $\mathrm{I}_{\mathrm{x}}=-i_{4}$,
MESH 1: $(4700+300) i_{1}-4700 i_{2} \quad=0 \quad[1]$
MESH 2: $\quad(4700+1700) i_{2}-4700 i_{1}-1700 i_{3}=0$
Since we have a current source on the perimeter of mesh 3, we do not require a KVL equation for that mesh. Instead, we may simply write

$$
\begin{array}{ccccc} 
& i_{3}=-0.03 v_{\pi} & {[3 a] \text { where }} & v_{\pi}=4700\left(i_{1}-i_{2}\right) \quad[3 b] \\
\text { MESH 4: } & 3000 i_{4}-3000 i_{3}+1 & =0 \tag{4}
\end{array}
$$

Simplifying and combining Eqs. $3 a$ and $3 b$,

$$
\begin{array}{rlc}
5000 i_{1}-4700 i_{2} & =0 \\
-4700 i_{1}+6400 i_{2}-1700 i_{3} & =0 \\
-141 i_{1}+141 i_{2}-i_{3} & =0 \\
-3000 i_{3}+3000 i_{4} & =-1
\end{array}
$$

Solving, we find that $i_{4}=-333.3 \mathrm{~mA}$, so

$$
\mathrm{I}_{\mathrm{x}}=333.3 \mu \mathrm{~A}
$$

(b) At node " $\pi$ " : $0.03 v_{\pi}=v_{\pi} / 300+v_{\pi} / 4700+v_{\pi} / 1700$

Solving, we find that $v_{\pi}=0$, therefore no current flows through the dependent source.
Hence, $\mathrm{I}_{\mathrm{x}}=333.3 \mu \mathrm{~A}$ as found in part (a).
(c) $\mathrm{V}_{\mathrm{s}} / \mathrm{I}_{\mathrm{x}}$ has units of resistance. It can be thought of as the resistance "seen" by the voltage source $\mathrm{V}_{\mathrm{s}} \ldots$... more on this in Chap. $5 . \ldots$.
40. We begin by naming each mesh and the three undefined voltage sources as shown below:


MESH 1: $\quad-\mathrm{V}_{\mathrm{Z}}+9 i_{1}-2 i_{2} \quad-7 i_{4}=0$
MESH 2: $-2 i_{1}+7 i_{2}-5 i_{3}=0$
MESH 3: $\quad \mathrm{V}_{\mathrm{x}} \quad-5 i_{2}+8 i_{3}-3 i_{4}=0$
MESH 4: $\quad \mathrm{V}_{\mathrm{y}}-7 i_{1} \quad-3 i_{3}+10 i_{4}=0$
Rearranging and setting $i_{1}-i_{2}=0, i_{2}-i_{3}=0, i_{1}-i_{4}=0$ and $i_{4}-i_{3}=0$,

$$
\begin{aligned}
9 i_{1}-2 i_{2}-7 i_{4} & =\mathrm{V}_{\mathrm{z}} \\
-2 i_{1}+7 i_{2}-5 i_{3} & =0 \\
-5 i_{2}+8 i_{3}-3 i_{4} & =-\mathrm{V}_{\mathrm{x}} \\
-7 i_{1}-3 i_{3}+10 i_{4} & =-\mathrm{V}_{\mathrm{y}}
\end{aligned}
$$

Since $i_{1}=i_{2}=i_{3}=i_{4}$, these equations produce:

$$
\begin{aligned}
\mathrm{V}_{\mathrm{z}} & =0 \\
0 & =0 \\
-\mathrm{V}_{\mathrm{x}} & =0 \\
-\mathrm{V}_{\mathrm{y}} & =0
\end{aligned}
$$

This is a unique solution. Therefore, the request that nonzero values be found cannot be satisfied.
41. The "supermesh" concept is not required (or helpful) in solving this problem, as there are no current sources shared between meshes. Starting with the left-most mesh and moving right, we define four clockwise mesh currents $i_{1}, i_{2}, i_{3}$ and $i_{4}$. By inspection, we see that $i_{1}=2 \mathrm{~mA}$.

MESH 2: $\quad-10+5000 i_{2}+4+1000 i_{3}=0$
MESH 3: $\quad-1000 i_{3}+6+10,000-10,000 i_{4}=0$
MESH 4: $\quad i_{4}=-0.5 i_{2}$
Reorganising, we find

$$
\begin{array}{rrrr}
5000 i_{2}+1000 i_{3} & =6 & {[1]} \\
9000 i_{3}-10,000 i_{4} & =-6 & {[2]} \\
0.5 i_{2} & +\quad i_{4} & =0 & {[3]}
\end{array}
$$

We could either subtitute Eq. [3] into Eq. [2] to reduce the number of equations, or simply go ahead and solve the system of Eqs. [1-3]. Either way, we find that

$$
i_{1}=2 \mathrm{~mA}, i_{2}=1.5 \mathrm{~mA}, i_{3}=-1.5 \mathrm{~mA} \text { and } i_{4}=-0.75 \mathrm{~mA} .
$$

The power generated by each source is:

$$
\begin{array}{ll}
\mathrm{P}_{2 \mathrm{~mA}}=5000\left(i_{1}-i_{2}\right)\left(i_{1}\right) & =5 \mathrm{~mW} \\
\mathrm{P}_{4 \mathrm{~V}}=4\left(-i_{2}\right) & =-6 \mathrm{~mW} \\
\mathrm{P}_{6 \mathrm{~V}}=6\left(-i_{3}\right) & =9 \mathrm{~mW} \\
\mathrm{P}_{\mathrm{dep}}=1000 i_{3}\left(i_{3}-i_{2}\right) & =4.5 \mathrm{~mW} \\
\mathrm{P}_{\mathrm{dep} I}=10,000\left(i_{3}-i_{4}\right)\left(0.5 i_{2}\right) & =-5.625 \mathrm{~mW}
\end{array}
$$

42. This circuit does not require the supermesh technique, as it does not contain any current sources. Redrawing the circuit so its planar nature and mesh structure are clear,


MESH 1: $\quad-20+i_{1}-i_{2}+2.5 i_{\mathrm{A}}=0$
MESH 2: $\quad 2 i_{2}+3 i_{2}+i_{2}-3 i_{3}-i_{1}=0$
MESH 3: $-2.5 i_{\mathrm{A}}+7 i_{3}-3 i_{2}=0$
Combining terms and making use of the fact that $i_{\mathrm{A}}=-i_{3}$,

$$
\begin{array}{rlr}
i_{1}-i_{2}-2.5 i_{3} & =20 & {[1]} \\
-i_{1}+6 i_{2}-3 i_{3} & =0 & {[2]} \\
-3 i_{2}+9.5 i_{3} & =0 & {[3]}
\end{array}
$$

Solving, $i_{1}=30.97 \mathrm{~A}, i_{2}=6.129 \mathrm{~A}$, and $i_{3}=1.936 \mathrm{~A}$. Since $i_{\mathrm{A}}=-i_{3}$,

$$
i_{\mathrm{A}}=-1.936 \mathrm{~A} .
$$

43. Define four mesh currents


By inspection, $i_{1}=-4.5 \mathrm{~A}$.
We form a supermesh with meshes 3 and 4 as defined above.
MESH 2:

$$
\begin{equation*}
2.2+3 i_{2}+4 i_{2}+5-4 i_{3}=0 \tag{1}
\end{equation*}
$$

SUPERMESH: $\quad 3 i_{4}+9 i_{4}-9 i_{1}+4 i_{3}-4 i_{2}+6 i_{3}+i_{3}-3=0$
Supermesh KCL equation: $\quad i_{4}-i_{3}=2$
Simplifying and combining terms, we may rewrite these three equations as:

$$
\begin{aligned}
7 i_{2}-4 i_{3} & = & -7.2 & {[1] } \\
-4 i_{2}+11 i_{3}+12 i_{4} & = & -37.5 & {[2] } \\
-i_{3}+i_{4} & = & 2 & {[3] }
\end{aligned}
$$

Solving, we find that $i_{2}=-2.839 \mathrm{~A}, i_{3}=-3.168 \mathrm{~A}$, and $i_{4}=-1.168 \mathrm{~A}$.
The power supplied by the 2.2-V source is then $2.2\left(i_{1}-i_{2}\right)=-3.654 \mathrm{~W}$.
44. We begin by defining six mesh currents as depicted below:


- We form a supermesh with meshes 1 and 2 since they share a current source.
- We form a second supermesh with meshes 3 and 4 since they also share a current source.

1, 2 Supermesh:
$(4700+1000+10,000) i_{1}-2200 i_{5}+(2200+1000+4700) i_{2}-1000 i_{3}=0$
3, 4 Supermesh:
$(4700+1000+2200) i_{3}-1000 i_{2}-2200 i_{6}+(4700+10,000+1000) i_{4}=0$
MESH 5:

$$
\begin{equation*}
(2200+4700) i_{5}-2200 i_{2}+3.2-1.5=0 \tag{2}
\end{equation*}
$$

MESH 6:

$$
\begin{equation*}
1.5+(4700+4700+2200) c-2200 i_{3}=0 \tag{3}
\end{equation*}
$$

1, 2 Supermesh KCL equation:

$$
\begin{align*}
& i_{1}-i_{2}=3 \times 10^{-3}  \tag{5}\\
& i_{4}-i_{3}=2 \times 10^{-3}
\end{align*}
$$

3, 4 Supermesh KCL equation:
We can simplify these equations prior to solution in several ways. Choosing to retain six equations,

$$
\begin{align*}
& 15,700 i_{1}+7900 i_{2}-1000 i_{3} \quad-2200 \mathrm{i} 5 \quad=0 \quad \text { [1] } \\
& -1000 i_{2}+7900 i_{3}+15,700 i_{4} \quad-2200 i_{6}=0 \quad \text { [2] } \\
& -2200 i_{2}+6900 i_{5} \quad=-1.7 \quad \text { [3] } \\
& -2200 i_{3} \quad+11,600 i_{6}=-1.5 \quad \text { [4] } \\
& i_{1}-i_{2} \quad=3 \times 10^{-3}  \tag{5}\\
& -i_{3}+i_{4} \quad=2 \times 10^{-3} \tag{6}
\end{align*}
$$

Solving, we find that $i_{4}=540.8 \mathrm{~mA}$. Thus, the voltage across the $2-\mathrm{mA}$ source is

$$
(4700+10,000+1000)\left(540.8 \times 10^{-6}\right)=8.491 \mathrm{~V}
$$

45. We define a mesh current $i_{\mathrm{a}}$ in the left-hand mesh, a mesh current $i_{1}$ in the top right mesh, and a mesh current $i_{2}$ in the bottom right mesh (all flowing clockwise).

The left-most mesh can be analysed separately to determine the controlling voltage $v_{\mathrm{a}}$, as KCL assures us that no current flows through either the $1-\Omega$ or $6-\Omega$ resistor.

Thus, $-1.8+3 i_{\mathrm{a}}-1.5+2 i_{\mathrm{a}}=0$, which may be solved to find $i_{\mathrm{a}}=0.66$ A. Hence, $v_{\mathrm{a}}=3 i_{\mathrm{a}}=1.98 \mathrm{~V}$.

Forming one supermesh from the remaining two meshes, we may write:

$$
-3+2.5 i_{1}+3 i_{2}+4 i_{2}=0
$$

and the supermesh KCL equation: $i_{2}-i_{1}=0.5 v_{\mathrm{a}}=0.5(1.98)=0.99$
Thus, we have two equations to solve:

$$
\begin{aligned}
& 2.5 i_{1}+7 i_{2}=3 \\
& -i_{1}+i_{2}=0.99
\end{aligned}
$$

Solving, we find that $i_{1}=-413.7 \mathrm{~mA}$ and the voltage across the $2.5-\Omega$ resistor (arbitrarily assuming the left terminal is the " + " reference) is $2.5 i_{1}=-1.034 \mathrm{~V}$.
46. There are only three meshes in this circuit, as the botton $22-\mathrm{m} \Omega$ resistor is not connected connected at its left terminal. Thus, we define three mesh currents, $i_{1}, i_{2}$, and $i_{3}$, beginning with the left-most mesh.

We next create a supermesh from meshes 1 and 2 (note that mesh 3 is independent, and can be analysed separately).

Thus,

$$
-11.8+10 \times 10^{-3} i_{1}+22 \times 10^{-3} i_{2}+10 \times 10^{-3} i_{2}+17 \times 10^{-3} i_{1}=0
$$

and applying KCL to obtain an equation containing the current source,

$$
i_{1}-i_{2}=100
$$

Combining terms and simplifying, we obtain

$$
\begin{aligned}
27 \times 10-3 i_{1}+32 \times 10^{-3} i_{2} & =11.8 \\
i_{1}-\quad i_{2} & =100
\end{aligned}
$$

Solving, we find that $i_{1}=254.2 \mathrm{~A}$ and $i_{2}=154.2 \mathrm{~A}$.
The final mesh current is easily found: $i_{3}=13 \times 10^{3} /(14+11.6+15)=320.2 \mathrm{~A}$.
47. MESH 1: $-7+i_{1}-i_{2}=0$

MESH 2: $\quad i_{2}-i_{1}+2 i_{2}+3 i_{2}-3 i_{3}=0$
MESH 3: $\quad 3 i_{3}-3 i_{2}+x i_{3}+2 i_{3}-7=0 \quad$ [3]
Grouping terms, we find that

$$
\begin{array}{lll}
i_{1}-i_{2} & =7 & {[1]} \\
-i_{1}+6 i_{2}-3 i_{3} & =0 & {[2]} \\
-3 i_{2}+(5+x) i_{3} & =7 & {[3]}
\end{array}
$$

This, unfortunately, is four unknowns but only three equations. However, we have not yet made use of the fact that we are trying to obtain $i_{2}=2.273 \mathrm{~A}$. Solving these "four" equations, we find that

$$
x=\left(7+3 i_{2}-5 i_{3}\right) / i_{3}=4.498 \Omega .
$$

48. We begin by redrawing the circuit as instructed, and define three mesh currents:


By inspection, $i_{3}=7 \mathrm{~A}$.
$\left.\begin{array}{llcc}\text { MESH 1: } & -7+i_{1}-i_{2}=0 & \text { or } & i_{1}-i_{2}=7 \\ \text { MESH 2: } & (1+2+3) i_{2}-i_{1}-3(7)=0 & \text { or } & -i_{1}+6 i_{2}=21\end{array}\right][2]$
There is no need for supermesh techniques for this situation, as the only current source lies on the outside perimeter of a mesh- it is not shared between meshes.
Solving, we find that $i_{1}=12.6 \mathrm{~A}, i_{2}=5.6 \mathrm{~A}$ and $i_{3}=7 \mathrm{~A}$.
49. (a) We are asked for a voltage, and have one current source and one voltage source. Nodal analysis is probably best then- the nodes can be named so that the desired voltage is a nodal voltage, or, at worst, we have one supernode equation to solve.

Name the top left node " 1 " and the top right node " x "; designate the bottom node as the reference terminal. Next, form a supernode with nodes " 1 " and "x."

At the supernode: $\quad 11=v_{1} / 2+v_{x} / 9$
and the KVL Eqn: $\quad v_{1}-v_{\mathrm{x}}=22$
Rearranging,

$$
\begin{equation*}
11(18)=9 v_{1}+2 v_{x} \tag{2}
\end{equation*}
$$

$22=v_{1}-v_{\mathrm{x}}$
Solving, $v_{\mathrm{x}}=0$
(b) We are asked for a voltage, and so may suspect that nodal analysis is preferrable; with two current sources and only one voltage source (easily dealt with using the supernode technique), nodal analysis does seem to have an edge over mesh analysis here.

Name the top left node "x," the top right node " $y$ " and designate the bottom node as the reference node. Forming a supernode from nodes " $x$ " and " $y$,"

At the supernode: $\quad 6+9=v_{x} / 10+v_{y} / 20 \quad$ [1]
and the KVL Eqn: $\quad v_{\mathrm{y}}-v_{\mathrm{x}}=12$
Rearranging, 15(20) $=2 v_{\mathrm{x}}+v_{\mathrm{y}} \quad$ [1]
and $\quad 12=-v_{\mathrm{x}}+v_{\mathrm{y}} \quad$ [2]
Solving, we find that $v_{\mathrm{x}}=96 \mathrm{~V}$.
(c) We are asked for a voltage, but would have to subtract two nodal voltages (not much harder than invoking Ohm's law). On the other hand, the dependent current source depends on the desired unknown, which would lead to the need for another equation if invoking mesh analysis. Trying nodal analysis,

$$
\begin{equation*}
0.1 v_{\mathrm{x}}=\left(v_{1}-50\right) / 2+v_{\mathrm{x}} / 4 \tag{1}
\end{equation*}
$$

referring to the circuit we see that $v_{\mathrm{x}}=v_{1}-100$. Rearranging so that we may eliminate $v_{1}$ in Eq. [1], we obtain $v_{1}=v_{\mathrm{x}}+100$. Thus, Eq. [1] becomes

$$
0.1 v_{\mathrm{x}}=\left(v_{\mathrm{x}}+100-50\right) / 2+v_{\mathrm{x}} / 4
$$

and a little algebra yields $v_{\mathrm{x}}=-38.46 \mathrm{~V}$.
50.

(a) We begin by noting that it is a voltage that is required; no current values are requested. This is a three-mesh circuit, or a four-node circuit, depending on your perspective. Either approach requires three equations.... Except that applying the supernode technique reduces the number of needed equations by one.
At the 1, 3 supernode:

$$
0=\left(v_{1}-80\right) / 10+\left(v_{1}-v_{3}\right) / 20+\left(v_{3}-v_{1}\right) / 20+v_{3} / 40+v_{3} / 30
$$

and

$$
v_{3}-v_{1}=30
$$

We simplify these two equations and collect terms, yielding

$$
\begin{gathered}
0.1 v_{1}+0.05833 v_{3}=8 \\
-v_{1}+v_{3}=30
\end{gathered}
$$

Solving, we find that $v_{3}=69.48 \mathrm{~V}$

Both ends of the resistor are connected to the supernode, so we could actually just ignore it..
(b) Mesh analysis would be straightforward, requiring 3 equations and a (trivial) application of Ohm's law to obtain the final answer. Nodal analysis, on the other hand, would require only two equations, and the desired voltage will be a nodal voltage.

At the b, c, d supernode: $\quad 0=\left(v_{b}-80\right) / 10+v_{\mathrm{d}} / 40+v_{\mathrm{d}} 30$
and: $\quad v_{\mathrm{d}}-v_{\mathrm{b}}=30 \quad v_{\mathrm{c}}-v_{\mathrm{d}}=9$
Simplify and collect terms: $\quad 0.1 v_{\mathrm{b}}+0.03333 v_{\mathrm{c}}+0.025 v_{\mathrm{d}}=80$

Solving, $v_{\mathrm{d}}\left(=v_{3}\right)=67.58 \mathrm{~V}$
(c) We are now faced with a dependent current source whose value depends on a mesh current. Mesh analysis in this situation requires 1 supermesh, 1 KCL equation and Ohm's law. Nodal analysis requires 1 supernode, 1 KVL equation, 1 other nodal equation, and one equation to express $i_{1}$ in terms of nodal voltages. Thus, mesh analysis has an edge here. Define the left mesh as " 1 ," the top mesh as " 2 ", and the bottom mesh as " 3 ."

Mesh 1:

$$
-80+10 i_{1}+20 i_{1}-20 i_{2}+30 i_{1}-30 i_{3}=0
$$

2, 3 supermesh: $\quad 20 i_{2}-20 i_{1}-30+40 i_{3}+30 i_{3}-30 i_{1}=0$
and:
$i_{2}-i_{3}=5 i_{1}$
Rewriting,
$60 i_{1}-20 i_{2}-30 i_{3}=80$
$-50 i_{1}+20 i_{2}+70 i_{3}=30$
$5 i_{1}-i_{2}+i_{3}=0$
Solving, $\quad i_{3}=4.727 \mathrm{~A}$
SO

$$
v_{3}=40 i_{3}=189 \mathrm{~V} .
$$

51. This circuit consists of 3 meshes, and no dependent sources. Therefore 3 simultaneous equations and 1 subtraction operation would be required to solve for the two desired currents. On the other hand, if we use nodal analysis, forming a supernode about the $30-\mathrm{V}$ source would lead to $5-1-1=3$ simulataneous equations as well, plus several subtraction and division operations to find the currents. Thus, mesh analysis has a slight edge here.

Define three clockwise mesh currents: $i_{\mathrm{a}}$ in the left-most mesh, $i_{\mathrm{b}}$ in the top right mesh, and $i_{\mathrm{c}}$ in the bottom right mesh. Then our mesh equations will be:

Mesh $a: \quad-80+(10+20+30) i_{\mathrm{a}}-20 i_{\mathrm{b}}-30 i_{\mathrm{c}}=0 \quad$ [1]
Mesh $b: \quad-30+(12+20) i_{\mathrm{b}}-12 i_{\mathrm{c}}-20 i_{\mathrm{a}}=0 \quad$ [2]
Mesh $c: \quad(12+40+30) i_{\mathrm{c}}-12 i_{\mathrm{b}}-30 i_{\mathrm{a}}=0 \quad$ [3]
Simplifying and collecting terms,

$$
\begin{array}{lll}
60 i_{\mathrm{a}}-20 i_{\mathrm{b}}-30 i_{\mathrm{c}}=80 & {[1]} \\
-20 i_{\mathrm{a}}+32 i_{\mathrm{b}}-12 i_{\mathrm{c}}=30 & {[2]} \\
-30 i_{\mathrm{a}}-12 i_{\mathrm{b}}+82 i_{\mathrm{c}}=0 & {[3]}
\end{array}
$$

Solving, we find that $i_{\mathrm{a}}=3.549 \mathrm{~A}, i_{\mathrm{b}}=3.854 \mathrm{~A}$, and $i_{\mathrm{c}}=1.863 \mathrm{~A}$. Thus,

$$
i_{1}=i_{\mathrm{a}}=3.549 \mathrm{~A} \quad \text { and } i_{2}=i_{\mathrm{a}}-i_{\mathrm{c}}=1.686 \mathrm{~A} .
$$

52. Approaching this problem using nodal analysis would require 3 separate nodal equations, plus one equation to deal with the dependent source, plus subtraction and division steps to actually find the current $i_{10}$. Mesh analysis, on the other hand, will require $2 \mathrm{mesh} /$ supermesh equations, 1 KCL equation, and one subtraction step to find $i_{10}$. Thus, mesh analysis has a clear edge. Define three clockwise mesh currents: $i_{1}$ in the bottom left mesh, $i_{2}$ in the top mesh, and $i_{3}$ in the bottom right mesh.

MESH 1: $\quad i_{1}=5 \mathrm{~mA}$ by inspection
SUPERMESH: $\quad i_{1}-i_{2}=0.4 i_{10}$
$i_{1}-i_{2}=0.4\left(i_{3}-i_{2}\right)$
$i_{1}-0.6 i_{2}-0.4 i_{3}=0$
MESH 3: $\quad-5000 i_{1}-10000 i_{2}+35000 i_{3}=0$
Simplify: $\quad 0.6 i_{2}+0.4 i_{3}=5 \times 10^{-3} \quad$ [2]

$$
-10000 i_{2}+35000 i_{3}=25
$$

Solving, we find $i_{2}=6.6 \mathrm{~mA}$ and $i_{3}=2.6 \mathrm{~mA}$. Since $i_{10}=i_{3}-i_{2}$, we find that

$$
i_{10}=-4 \mathrm{~mA} .
$$

53. For this circuit problem, nodal analysis will require 3 simultaneous nodal equations, then subtraction/ division steps to obtain the desired currents. Mesh analysis requires 1 mesh equation, 1 supermesh equation, 2 simple KCL equations and one subtraction step to determine the currents. If either technique has an edge in this situation, it's probably mesh analysis. Thus, define four clockwise mesh equations: $i_{\mathrm{a}}$ in the bottom left mesh, $i_{\mathrm{b}}$ in the top left mesh, $i_{\mathrm{c}}$ in the top right mesh, and $i_{\mathrm{d}}$ in the bottom right mesh.

At the $a, b, c$ supermesh: $\quad-100+6 i_{\mathrm{a}}+20 i_{\mathrm{b}}+4 i_{\mathrm{c}}+10 i_{\mathrm{c}}-10 i_{\mathrm{d}}=0$
Mesh d:

$$
\begin{equation*}
100+10 \mathrm{id}-10 i_{\mathrm{c}}+24 i_{\mathrm{d}}=0 \tag{1}
\end{equation*}
$$

KCL:
$-i_{\mathrm{a}}+i_{\mathrm{b}}=2$
and

$$
\begin{equation*}
-i_{\mathrm{b}}+i_{\mathrm{c}}=3 i_{3}=3 i_{\mathrm{a}} \tag{3}
\end{equation*}
$$

Collecting terms \& simplifying,

$$
\begin{aligned}
6 i_{\mathrm{a}}+20 i_{\mathrm{b}}+14 i_{\mathrm{c}}-10 i_{\mathrm{d}} & =100 & & {[1] } \\
-10 i_{\mathrm{c}}+34 i_{\mathrm{d}} & =-100 & & {[2] } \\
-i_{\mathrm{a}}+i_{\mathrm{b}} & & 2 & {[3] } \\
-3 i_{\mathrm{a}}-i_{\mathrm{b}}+i_{\mathrm{c}} & & 0 & {[4] }
\end{aligned}
$$

Solving,
$i_{\mathrm{a}}=0.1206 \mathrm{~A}, i_{\mathrm{b}}=2.121 \mathrm{~A}, i_{\mathrm{c}}=2.482 \mathrm{~A}$, and $i_{\mathrm{d}}=-2.211 \mathrm{~A}$. Thus,

$$
i_{3}=i_{\mathrm{a}}=120.6 \mathrm{~mA} \text { and } i_{10}=i_{\mathrm{c}}-i_{\mathrm{d}}=4.693 \mathrm{~A} .
$$

54. With 7 nodes in this circuit, nodal analysis will require the solution of three simultaneous nodal equations (assuming we make use of the supernode technique) and one KVL equation. Mesh analysis will require the solution of three simultaneous mesh equations (one mesh current can be found by inspection), plus several subtraction and multiplication operations to finally determine the voltage at the central node. Either will probably require a comparable amount of algebraic manoeuvres, so we go with nodal analysis, as the desired unknown is a direct result of solving the simultaneous equations. Define the nodes as:


NODE 1: $\quad-2 \times 10^{-3}=\left(v_{1}-1.3\right) / 1.8 \times 10^{3} \quad \rightarrow \quad v_{1}=-2.84 \mathrm{~V}$.
2, 4 Supernode:
$2.3 \times 10^{-3}=\left(v_{2}-v_{5}\right) / 1 \times 10^{3}+\left(v_{4}-1.3\right) / 7.3 \times 10^{3}+\left(v_{4}-v_{5}\right) / 1.3 \times 10^{3}+v_{4} / 1.5 \times 10^{3}$
KVL equation: $\quad-v_{2}+v_{4}=5.2$
Node 5: $\quad 0=\left(v_{5}-v_{2}\right) / 1 \times 10^{3}+\left(v_{5}-v_{4}\right) / 1.3 \times 10^{3}+\left(v_{5}-2.6\right) / 6.3 \times 10^{3}$
Simplifying and collecting terms,

$$
\begin{array}{rlrl}
14.235 \mathrm{v}_{2}+22.39 \mathrm{v}_{4}-25.185 \mathrm{v}_{5} & =35.275 \\
-\mathrm{v}_{2}+ & \mathrm{v}_{4} & =5.2 \\
-8.19 \mathrm{v}_{2} & =6.3 \mathrm{v}_{4}+15.79 \mathrm{v}_{5} & =3.38
\end{array}
$$

Solving, we find the voltage at the central node is $v_{4}=3.460 \mathrm{~V}$.
55. Mesh analysis yields current values directly, so use that approach. We therefore define four clockwise mesh currents, starting with $i_{1}$ in the left-most mesh, then $i_{2}, i_{3}$ and $i_{4}$ moving towards the right.

Mesh 1: $\quad-0.8 i_{x}+(2+5) i_{1}-5 i_{2}=0$
Mesh 2: $\quad i_{2}=1$ A by inspection
Mesh 3: $\quad(3+4) i_{3}-3(1)-4\left(i_{4}\right)=0$
Mesh 4: $\quad(4+3) i_{4}-4 i_{3}-5=0$ [4]

Simplify and collect terms, noting that $i_{\mathrm{x}}=i_{1}-i_{2}=i_{1}-1$
$-0.8\left(i_{1}-1\right)+7 i_{1}-5(1)=0$ yields $i_{1}=677.4 \mathrm{~mA}$
Thus, [3] and [4] become: $7 i_{3}-4 i_{4}=3$
$-4 i_{3}+7 i_{4}=5$
Solving, we find that $i_{3}=1.242 \mathrm{~A}$ and $i_{4}=1.424 \mathrm{~A}$. A map of individual branch currents can now be drawn:

56. If we choose to perform mesh analysis, we require 2 simultaneous equations (there are four meshes, but one mesh current is known, and we can employ the supermesh technique around the left two meshes). In order to find the voltage across the $2-\mathrm{mA}$ source we will need to write a KVL equation, however. Using nodal analysis is less desirable in this case, as there will be a large number of nodal equations needed. Thus, we define four clockwise mesh currents $i_{1}, i_{2}, i_{3}$ and $i_{4}$ starting with the leftmost mesh and moving towards the right of the circuit.

At the 1,2 supermesh: $\quad 2000 i_{1}+6000 i_{2}-3+5000 i_{2}=0$
and

$$
\begin{equation*}
i_{1}-i_{2}=2 \times 10^{-3} \tag{1}
\end{equation*}
$$

by inspection, $i_{4}=-1 \mathrm{~mA}$. However, this as well as any equation for mesh four are unnecessary: we already have two equations in two unknowns and $i_{1}$ and $i_{2}$ are sufficient to enable us to find the voltage across the current source.

Simplifying, we obtain $\quad 2000 i_{1}+11000 i_{2}=3 \quad$ [1]

$$
1000 i_{1}-1000 i_{2}=2
$$

Solving, $i_{1}=1.923 \mathrm{~mA}$ and $i_{2}=-76.92 \mu \mathrm{~A}$.
Thus, the voltage across the $2-\mathrm{mA}$ source ("+" reference at the top of the source) is

$$
v=-2000 i_{1}-6000\left(i_{1}-i_{2}\right)=-15.85 \mathrm{~V} .
$$

57. Nodal analysis will require 2 nodal equations (one being a "supernode" equation), 1 KVL equation, and subtraction/division operations to obtain the desired current. Mesh analysis simply requires 2 "supermesh" equations and 2 KCL equations, with the desired current being a mesh current. Thus, we define four clockwise mesh currents $i_{\mathrm{a}}, i_{\mathrm{b}}, i_{\mathrm{c}}, i_{\mathrm{d}}$ starting with the left-most mesh and proceeding to the right of the circuit.

At the $a, b$ supermesh: $\quad-5+2 i_{\mathrm{a}}+2 i_{\mathrm{b}}+3 i_{\mathrm{b}}-3 i_{\mathrm{c}}=0$
At the $c, d$ supermesh: $\quad 3 i_{\mathrm{c}}-3 i_{\mathrm{b}}+1+4 i_{\mathrm{d}}=0$
and

$$
\begin{array}{ll}
i_{\mathrm{a}}-i_{\mathrm{b}}=3 & {[3]}  \tag{2}\\
i_{\mathrm{c}}-i_{\mathrm{d}}=2 & {[4]}
\end{array}
$$

Simplifying and collecting terms, we obtain

$$
\begin{aligned}
2 i_{\mathrm{a}}+5 i_{\mathrm{b}}-3 i_{\mathrm{c}} & =5 & & {[1] } \\
-3 i_{\mathrm{b}}+3 i_{\mathrm{c}}+4 i_{\mathrm{d}} & =-1 & & {[2] } \\
i_{\mathrm{a}}-i_{\mathrm{b}} & & =3 & \\
& & i_{\mathrm{c}}-i_{\mathrm{d}} & =2
\end{aligned}
$$

Solving, we find $i_{\mathrm{a}}=3.35 \mathrm{~A}, i_{\mathrm{b}}=350 \mathrm{~mA}, i_{\mathrm{c}}=1.15 \mathrm{~A}$, and $i_{\mathrm{d}}=-850 \mathrm{~mA}$. As $i_{1}=i_{\mathrm{b}}$,

$$
i_{1}=350 \mathrm{~mA} .
$$

58. Define a voltage $v_{x}$ at the top node of the current source $\mathrm{I}_{2}$, and a clockwise mesh current $i_{\mathrm{b}}$ in the right-most mesh.

We want 6 W dissipated in the $6-\Omega$ resistor, which leads to the requirement $i_{\mathrm{b}}=1 \mathrm{~A}$. Applying nodal analysis to the circuit,
$\mathrm{I}_{1}+\mathrm{I}_{2}=\left(v_{\mathrm{x}}-v_{1}\right) / 6=1$
so our requirement is $\mathrm{I}_{1}+\mathrm{I}_{2}=1$. There is no constraint on the value of $v_{1}$ other than we are told to select a nonzero value.

Thus, we choose $\mathrm{I}_{1}=\mathrm{I}_{2}=500 \mathrm{~mA}$ and $v_{1}=3.1415 \mathrm{~V}$.
59. Inserting the new $2-\mathrm{V}$ source with "+" reference at the bottom, and the new $7-\mathrm{mA}$ source with the arrow pointing down, we define four clockwise mesh currents $i_{1}, i_{2}, i_{3}$, $i_{4}$ starting with the left-most mesh and proceeding towards the right of the circuit.

Mesh 1:

$$
\begin{equation*}
(2000+1000+5000) i_{1}-6000 i_{2}-2=0 \tag{1}
\end{equation*}
$$

2, 3 Supermesh:
$2+(5000+5000+1000+6000) i_{2}-6000 i_{1}+(3000+4000+5000) i_{3}-5000 i_{4}$ $=0 \quad[2]$
and

$$
\begin{equation*}
i_{2}-i_{3}=7 \times 10^{-3} \tag{3}
\end{equation*}
$$

Mesh 4:

$$
\begin{equation*}
i_{4}=-1 \mathrm{~mA} \text { by inspection } \tag{4}
\end{equation*}
$$

Simplifying and combining terms,

$$
\begin{array}{rlll}
8000 i_{1}-6000 i_{2} & & 2 & {[1]} \\
1000 i_{2}-1000 i_{3} & & =7 & {[4]} \\
-6000 i_{1}+17000 i_{2}+12000 i_{3} & =-7 & {[2]}
\end{array}
$$

Solving, we find that

$$
i_{1}=2.653 \mathrm{~A}, i_{2}=3.204 \mathrm{~A}, i_{3}=-3.796 \mathrm{~A}, i_{4}=-1 \mathrm{~mA}
$$

60. This circuit is easily analyzed by mesh analysis; it's planar, and after combining the 2A and 3 A sources into a single 1 A source, supermesh analysis is simple.

First, define clockwise mesh currents $i_{x}, i_{1}, i_{2}$ and $i_{3}$ starting from the left-most mesh and moving to the right. Next, combine the 2 A and 3 A sources temporarily into a 1 A source, arrow pointing upwards. Then, define four nodal voltages, $\mathrm{V}_{1}, \mathrm{~V}_{2}, \mathrm{~V}_{3}$ and $\mathrm{V}_{4}$ moving from left to right along the top of the circuit.

At the left-most mesh, $i_{x}=-5 i_{1}$
For the supermesh, we can write $4 i_{1}-2 i_{x}+2+2 i_{3}=0$
and the corresponding KCL equation: $i_{3}-i_{1}=1$
Substituting Eq. [1] into Eq. [2] and simplifying,
$14 i_{1}+2 i_{3}=-2$
$-i_{1}+i_{3}=1$
Solving, $i_{1}=-250 \mathrm{~mA}$ and $i_{3}=750 \mathrm{~mA}$.
Then, $i_{\mathrm{x}}=-5 i_{1}=1.35 \mathrm{~A}$ and $i_{2}=i_{1}-2=-2.25 \mathrm{~A}$
Nodal voltages are straightfoward to find, then:

$$
\begin{array}{|l}
\hline \mathrm{V}_{4}=2 i_{3}=1.5 \mathrm{~V} \\
\mathrm{~V}_{3}=2+\mathrm{V}_{4}=3.5 \mathrm{~V} \\
\mathrm{~V}_{2}-\mathrm{V}_{3}=2 i_{1} \text { or } \mathrm{V}_{2}=2 i_{1}+\mathrm{V}_{3}=3 \mathrm{~V} \\
\mathrm{~V}_{1}-\mathrm{V}_{2}=2 i_{\mathrm{x}} \text { or } \mathrm{V}_{1}=2 i_{\mathrm{x}}+\mathrm{V}_{2}=5.5 \mathrm{~V}
\end{array}
$$

## 61. Hand analysis:

Define three clockwise mesh currents: $i_{1}$ in the bottom left mesh, $i_{2}$ in the top mesh, and $i_{3}$ in the bottom right mesh.

MESH 1: $\quad i_{1}=5 \mathrm{~mA}$ by inspection
SUPERMESH: $\quad i_{1}-i_{2}=0.4 i_{10}$
$i_{1}-i_{2}=0.4\left(i_{3}-i_{2}\right)$
$i_{1}-0.6 i_{2}-0.4 i_{3}=0$
MESH 3: $\quad-5000 i_{1}-10000 i_{2}+35000 i_{3}=0$
Simplify: $\quad 0.6 i_{2}+0.4 i_{3}=5 \times 10^{-3} \quad$ [2] $-10000 i_{2}+35000 i_{3}=25 \quad[3]$

Solving, we find $i_{2}=6.6 \mathrm{~mA}$ and $i_{3}=2.6 \mathrm{~mA}$. Since $i_{10}=i_{3}-i_{2}$, we find that

$$
i_{10}=-4 \mathrm{~mA} .
$$

PSpice simulation results:


Summary: The current entering the right-hand node of the $10-\mathrm{k} \Omega$ resistor R2 is equal to 4.000 mA . Since this current is $-i_{10}, i_{10}=-4.000 \mathrm{~mA}$ as found by hand.

## 62. Hand analysis:

Define the nodes as:


NODE 1:

$$
-2 \times 10^{-3}=\left(v_{1}-1.3\right) / 1.8 \times 10^{3} \quad \rightarrow v_{1}=-2.84 \mathrm{~V}
$$

2, 4 Supernode:
$2.3 \times 10^{-3}=\left(v_{2}-v_{5}\right) / 1 \times 10^{3}+\left(v_{4}-1.3\right) / 7.3 \times 10^{3}+\left(v_{4}-v_{5}\right) / 1.3 \times 10^{3}+v_{4} / 1.5 \times 10^{3}$
KVL equation: $\quad-v_{2}+v_{4}=5.2$
Node 5: $\quad 0=\left(v_{5}-v_{2}\right) / 1 \times 10^{3}+\left(v_{5}-v_{4}\right) / 1.3 \times 10^{3}+\left(v_{5}-2.6\right) / 6.3 \times 10^{3}$
Simplifying and collecting terms,

$$
\begin{array}{rlrl}
14.235 \mathrm{v}_{2}+22.39 \mathrm{v}_{4}-25.185 \mathrm{v}_{5} & =35.275 & {[1]} \\
-\mathrm{v}_{2}+ & \mathrm{v}_{4} & =5.2 & {[2]} \\
-8.19 \mathrm{v}_{2} & -6.3 \mathrm{v}_{4}+15.79 \mathrm{v}_{5} & =3.38
\end{array}
$$

Solving, we find the voltage at the central node is $v_{4}=3.460 \mathrm{~V}$.

## PSpice simulation results:



Summary: The voltage at the center node is found to be 3.460 V , which is in agreement with our hand calculation.

## 63. Hand analysis:

At the 1,2 supermesh:

$$
\begin{equation*}
2000 i_{1}+6000 i_{2}-3+5000 i_{2}=0 \tag{1}
\end{equation*}
$$

and

$$
\begin{equation*}
i_{1}-i_{2}=2 \times 10^{-3} \tag{2}
\end{equation*}
$$

by inspection, $i_{4}=-1 \mathrm{~mA}$. However, this as well as any equation for mesh four are unnecessary: we already have two equations in two unknowns and $i_{1}$ and $i_{2}$ are sufficient to enable us to find the voltage across the current source.

Simplifying, we obtain

$$
\begin{align*}
& 2000 i_{1}+11000 i_{2}=3  \tag{1}\\
& 1000 i_{1}-1000 i_{2}=2 \tag{2}
\end{align*}
$$

Solving, $i_{1}=1.923 \mathrm{~mA}$ and $i_{2}=-76.92 \mu \mathrm{~A}$.
Thus, the voltage across the $2-\mathrm{mA}$ source ("+" reference at the top of the source) is

$$
v=-2000 i_{1}-6000\left(i_{1}-i_{2}\right)=-15.85 \mathrm{~V} .
$$

## PSpice simulation results:



Summary: Again arbitrarily selecting the " + " reference as the top node of the $2-\mathrm{mA}$ current source, we find the voltage across it is $-5.846-10=-15.846 \mathrm{~V}$, in agreement with our hand calculation.

## 64. Hand analysis:

Define a voltage $v_{x}$ at the top node of the current source $I_{2}$, and a clockwise mesh current $i_{\mathrm{b}}$ in the right-most mesh.

We want 6 W dissipated in the $6-\Omega$ resistor, which leads to the requirement $i_{\mathrm{b}}=1 \mathrm{~A}$. Applying nodal analysis to the circuit,
$\mathrm{I}_{1}+\mathrm{I}_{2}=\left(v_{\mathrm{x}}-v_{1}\right) / 6=1$
so our requirement is $\mathrm{I}_{1}+\mathrm{I}_{2}=1$. There is no constraint on the value of $v_{1}$ other than we are told to select a nonzero value.

Thus, we choose $\mathrm{I}_{1}=\mathrm{I}_{2}=500 \mathrm{~mA}$ and $v_{1}=3.1415 \mathrm{~V}$.

## PSpice simulation results:



Summary: We see from the labeled schematic above that our choice for $\mathrm{I}_{1}, \mathrm{I}_{2}$ and $V_{1}$ lead to 1 A through the $6-\Omega$ resistor, or 6 W dissipated in that resistor, as desired.

## 65. Hand analysis:

Define node 1 as the top left node, and node 2 as the node joining the three $2-\Omega$ resistors. Place the "+" reference terminal of the $2-\mathrm{V}$ source at the right. The rightmost $2-\Omega$ resistor has therefore been shorted out. Applying nodal analysis then,

Node 1: $\quad-5 i_{1}=\left(v_{1}-v_{2}\right) / 2$
Node 2: $\quad 0=\left(v_{2}-v_{1}\right) / 2+v_{2} / 2+\left(v_{2}-2\right) / 2$
and,

$$
\begin{equation*}
i_{1}=\left(v_{2}-2\right) / 2 \tag{2}
\end{equation*}
$$

Simplifying and collecting terms,

$$
\begin{align*}
& v_{1}+v_{2}=10 \\
& -v_{1}+3 v_{2}=2 \tag{2}
\end{align*}
$$

Solving, we find that $v_{1}=3.143 \mathrm{~V}$ and $v_{2}=1.714 \mathrm{~V}$.
Defining clockwise mesh currents $\mathrm{i}_{\mathrm{a}}, \mathrm{i}_{\mathrm{b}}, \mathrm{i}_{\mathrm{c}}, \mathrm{i}_{\mathrm{d}}$ starting with the left-most mesh and proceeding right, we may easily determine that

$$
\begin{aligned}
& i_{\mathrm{a}}=-5 i_{1}=714.3 \mathrm{~mA} \\
& i_{\mathrm{b}}=-142.9 \mathrm{~mA} \\
& i_{\mathrm{c}}=i_{1}-2=-2.143 \mathrm{~A} \\
& i_{\mathrm{d}}=3+i_{\mathrm{c}}=857.1 \mathrm{~mA}
\end{aligned}
$$

## PSpice simulation results:



Summary: The simulation results agree with the hand calculations.
66. (a) One possible circuit configuration of many that would satisfy the requirements:


At node 1: $\quad-3=\left(v_{1}-5\right) / 100+\left(v_{1}-v 2\right) / 50$
At node 2: $2 v_{\mathrm{x}}=\left(\mathrm{v} 2-v_{1}\right) / 50+v_{2} / 30$
and,

$$
\begin{equation*}
v_{\mathrm{x}}=5-v_{1} \tag{2}
\end{equation*}
$$

Simplifying and collecting terms,

$$
\begin{align*}
& 150 v_{1}-100 v_{2}=-14750  \tag{1}\\
& 2970 v_{1}+80 v_{2}=15000 \tag{2}
\end{align*}
$$

Solving, we find that $v_{1}=1.036 \mathrm{~V}$ and $v_{2}=149.1 \mathrm{~V}$.
The current through the $100-\Omega$ resistor is simply $\left(5-v_{1}\right) / 100=39.64 \mathrm{~mA}$
The current through the $50-\Omega$ resistor is $\left(v_{1}-v_{2}\right) / 50=-2.961 \mathrm{~A}$, and the current through the $20-\Omega$ and $10-\Omega$ series combination is $v_{2} / 30=4.97 \mathrm{~A}$.
Finally, the dependent source generates a current of $2 v_{x}=7.928 \mathrm{~A}$.
(b) PSpice simulation results


Summary: The simulated results agree with the hand calculations.
67. One possible solution of many:


Choose R so that $3 \mathrm{R}=5$; then the voltage across the current source will be 5 V , and so will the voltage across the resistor R .
$\mathrm{R}=5 / 3 \Omega$. To construct this from $1-\Omega$ resistors, note that

$$
5 / 3 \Omega=1 \Omega+2 / 3 \Omega=1 \Omega+1 \Omega\|1 \Omega\| 1 \Omega+1 \Omega\|1 \Omega\| 1 \Omega
$$

```
* Solution to Problem 4.57
.OP
V110 DC 10
I1 04 DC 3
R1121
R2 2 3 1
R3 2 31
R4231
R5 341
R6 341
R7 341
.END
```

```
**** SMALL SIGNAL BIAS SOLUTION TEMPERATURE = 27.000 DEG C
******************************************************************************
NODE VOLTAGE NODE VOLTAGE NODE VOLTAGE NODE VOLTAGE
( 1) 10.0000 ( 2) 7.0000 ( 3) 6.0000 ( 4) 5.0000
    VOLTAGE SOURCE CURRENTS
    NAME CURRENT
    V1 -3.000E+00
```

68. We first name each node, resistor and voltage source:


We next write an appropriate input deck for SPICE:

```
* Solution to Problem 4.58
.OP
V11 0 DC 20
R1122
R2 }20
R3234
R42410
R5 3 }0
R6346
R7 3511
R8407
R9458
R10509
.END
```

And obtain the following output:


We see from this simulation result that the voltage $v_{5}=2.847 \mathrm{~V}$.
69. One possible solution of many:


Verify:

$$
\begin{aligned}
& v_{1}=9(4 / 9)=4 \mathrm{~V} \\
& v_{2}=9(3 / 9)=3 \mathrm{~V} \\
& v_{3}=9(2 / 9)=2 \mathrm{~V}
\end{aligned}
$$

SPICE INPUT DECK:

```
* Solution to Problem 4.59
.OP
V110 DC 9
R1125
R2 231
R3 341
R4451
R5 501
.END
```



PROPRIETARY MATERIAL. © 2007 The McGraw-Hill Companies, Inc. Limited distribution permitted only to teachers and educators for course preparation. If you are a student using this Manual, you are using it without permission.
70. (a) If only two bulbs are not lit (and thinking of each bulb as a resistor), the bulbs must be in parallel otherwise, the burned out bulbs, acting as short circuits, would prevent current from flowing to the "good" bulbs.
(b) In a parallel connected circuit, each bulb "sees" 115 VAC. Therefore, the individual bulb current is $1 \mathrm{~W} / 115 \mathrm{~V}=8.696 \mathrm{~mA}$. The resistance of each "good" bulb is $\mathrm{V} / \mathrm{I}=13.22 \mathrm{k} \Omega$. A simplified, electrically-equivalent model for this circuit would be a 115 VAC source connected in parallel to a resistor $\mathrm{R}_{\text {eq }}$ such that
$1 / \mathrm{R}_{\mathrm{eq}}=1 / 13.22 \times 10^{3}+1 / 13.22 \times 10^{3}+\ldots .+1 / 13.22 \times 10^{3} \quad$ ( $400-2=398$ terms $)$ or $\mathrm{R}_{\mathrm{eq}}=33.22 \Omega$. We expect the source to provide 398 W .

```
* Solution to Problem 4.60
.OP
V110 AC 11560
R11033.22
.AC LIN 1 }606
.PRINT AC VM(1)IM(V1)
.END
```

```
**** 07/29/01 21:09:32 *********** Evaluation PSpice (Nov 1999) **************
* Solution to Problem 4.60
**** SMALL SIGNAL BIAS SOLUTION TEMPERATURE = 27.000 DEG C
**********************************************************************************
NODE VOLTAGE NODE VOLTAGE NODE VOLTAGE NODE VOLTAGE
( 1) 0.0000
    VOLTAGE SOURCE CURRENTS
    NAME CURRENT
    V1 0.000E+00
    TOTAL POWER DISSIPATION 0.00E+00 WATTS
**** 07/29/01 21:09:32 *********** Evaluation PSpice (Nov 1999)
* Solution to Problem 4.60
**** AC ANALYSIS TEMPERATURE = 27.000 DEG C
FREQ VM(1) IM(V1)
6.000E+01 1.150E+02 3.462E+00
```

(c) The inherent series resistance of the wire connections leads to a voltage drop which increases the further one is from the voltage source. Thus, the furthest bulbs actually have less than 115 VAC across them, so they draw slightly less current and glow more dimly.

